LV8772

ON Semiconductor®

http://onsemi.com

Bi-CMOS LSI

PWM Constant-Current Control Stepping Motor Driver

Overview

The LV8772 is a stepping motor driver, which is capable of micro-step drive and supports 4W 1-2 phase excitation. It is stepping motors used in office equipment and amusement applications.

Features

- Low on resistance (upper side : 0.3Ω ; lower side : 0.25Ω ; total of upper and lower : 0.55Ω ; Ta = 25°C, I_O = 2.5A)
- Excitation mode can be set to 2-phase, 1-2 phase, W1-2 phase , or 4W1-2 phase
- BiCDMOS process IC
- Motor current selectable in four steps
- Unusual condition warning output pins
- Excitation step proceeds only by step signal input
- Output short-circuit protection circuit incorporated
- No control power supply required

Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	VM max		36	V
Output peak current	I_O peak	tw ≤ 10ms, duty 20%	3.0	A
Output current	I_O max		2.5	A
Logic input voltage	V_{IN}		-0.3 to +6	V
MONI/EMO input voltage	Vmoni/Vemo		-0.3 to +6	V
Allowable power dissipation	Pd max1	1 unit	3.0	W
	Pd max2	*	5.4	W
Operating temperature	Topr		-20 to +85	°C
Storage temperature	Tstg		-55 to +150	°C

* Specified circuit board : 90.0mm×90.0mm×1.6mm, glass epoxy 2-layer board, with backside mounting.

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.

Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details.

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Allowable Operating Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage range	VM		9 to 32	V
Logic input voltage	V_{IN}		0 to 5.5	V
VREF input voltage range	VREF		0 to 3	V

Electrical Characteristics at Ta = 25°C, VM = 24V, VREF = 1.5V

Parameter		Symbol	Conditions	Ratings min	Ratings typ	Ratings max	Unit
Standby mode current drain		IMst	ST = "L"		100	400	µA
Current drain		IM	ST = "H", with no load		3.2	5	mA
VREG5 output voltage		Vreg5	I_O = -1mA	4.5	5	5.5	V
Thermal shutdown temperature		TSD	Design guarantee	150	180	200	°C
Thermal hysteresis width		ΔTSD	Design guarantee		40		°C
Motor driver							
Output on resistance		Ronu	I_O = 2.5A, Upper-side on resistance		0.3	0.4	Ω
		Rond	I_O = 2.5A, Lower-side on resistance		0.25	0.33	Ω
Output leakage current		I_Oleak				50	µA
Diode forward voltage		VD	ID = -2.5A		1.2	1.4	V
Logic pin input current		$I_{IN}L$	V_{IN} = 0.8V	4	8	12	µA
		$I_{IN}H$	V_{IN} = 5V	30	50	70	µA
Logic high-level input voltage		$V_{IN}H$		2.0			V
Logic low-level input voltage		$V_{IN}L$				0.8	V
Current setting comparator threshold voltage (current step switching)	4W1-2-phase drive	Vtdac0_4W	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
		Vtdac1_4W	Step 1 (Initial state+1)	0.291	0.3	0.309	V
		Vtdac2_4W	Step 2 (Initial state+2)	0.285	0.294	0.303	V
		Vtdac3_4W	Step 3 (Initial state+3)	0.279	0.288	0.297	V
		Vtdac4_4W	Step 4 (Initial state+4)	0.267	0.276	0.285	V
		Vtdac5_4W	Step 5 (Initial state+5)	0.255	0.264	0.273	V
		Vtdac6_4W	Step 6 (Initial state+6)	0.240	0.249	0.258	V
		Vtdac7_4W	Step 7 (Initial state+7)	0.222	0.231	0.240	V
		Vtdac8_4W	Step 8 (Initial state+8)	0.201	0.210	0.219	V
		Vtdac9_4W	Step 9 (Initial state+9)	0.180	0.189	0.198	V
		Vtdac10_4W	Step 10 (Initial state+10)	0.157	0.165	0.173	V
		Vtdac11_4W	Step 11 (Initial state+11)	0.134	0.141	0.148	V
		Vtdac12_4W	Step 12 (Initial state+12)	0.107	0.114	0.121	V
		Vtdac13_4W	Step 13 (Initial state+13)	0.080	0.087	0.094	V
		Vtdac14_4W	Step 14 (Initial state+14)	0.053	0.060	0.067	V
		Vtdac15_4W	Step 15 (Initial state+15)	0.023	0.030	0.037	V
	W1-2-phase drive	Vtdac0_W	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
		Vtdac4_W	Step 4 (Initial state+1)	0.267	0.276	0.285	V
		Vtdac8_W	Step 8 (Initial state+2)	0.201	0.21	0.219	V
		Vtdac12_W	Step 12 (Initial state+3)	0.107	0.114	0.121	V
	1-2 phase drive	Vtdac0_H	Step 0 (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
		Vtdac8_H	Step 8 (Initial state+1)	0.201	0.21	0.219	V
	2 phase drive	Vtdac8_F	Step 8' (When initialized : channel 1 comparator level)	0.291	0.3	0.309	V
Current setting comparator threshold voltage (current attenuation rate switching)		Vtatt00	ATT1 = L, ATT2 = L	0.291	0.3	0.309	V
		Vtatt01	ATT1 = H, ATT2 = L	0.232	0.24	0.248	V
		Vtatt10	ATT1 = L, ATT2 = H	0.143	0.15	0.157	V
		Vtatt11	ATT1 = H, ATT2 = H	0.053	0.06	0.067	V
Chopping frequency		Fchop	Cchop = 180pF	45	55	65	kHz
CHOP pin charge/discharge current		Ichop		7	10	13	µA

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Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Chopping oscillation circuit threshold voltage	Vtup		0.8	1	1.2	V
	Vtdown		0.4	0.5	0.6	V
VREF pin input current	Iref	VREF = 1.5V	-0.5			µA
MONI pin saturation voltage	Vsatmon	Imoni = 1mA			400	mV
Charge pump						
VG output voltage	VG		28	28.7	29.8	V
Rise time	tONG	VG = 0.1µF, CP1-CP2 = 0.1µF, ST = "H" →VG = VM+4V		200	500	µS
Oscillator frequency	Fosc		90	125	150	kHz
Output short-circuit protection						
EMO pin saturation voltage	Vsatemo	Iemo = 1mA			400	mV

Package Dimensions

unit : mm (typ)

3147C

DIP28H(500mil)

Pd max - Ta

Block Diagram

Pin Assignment
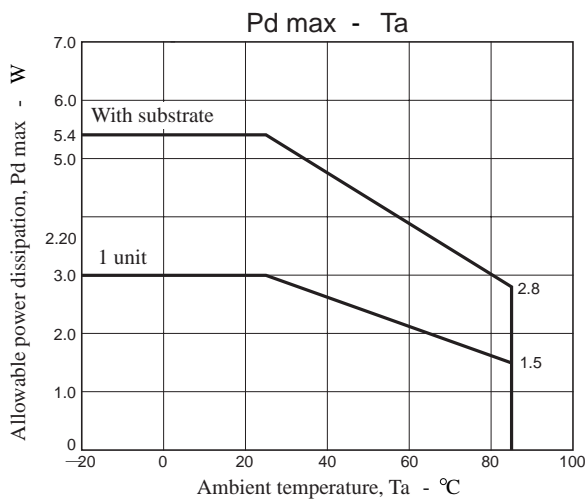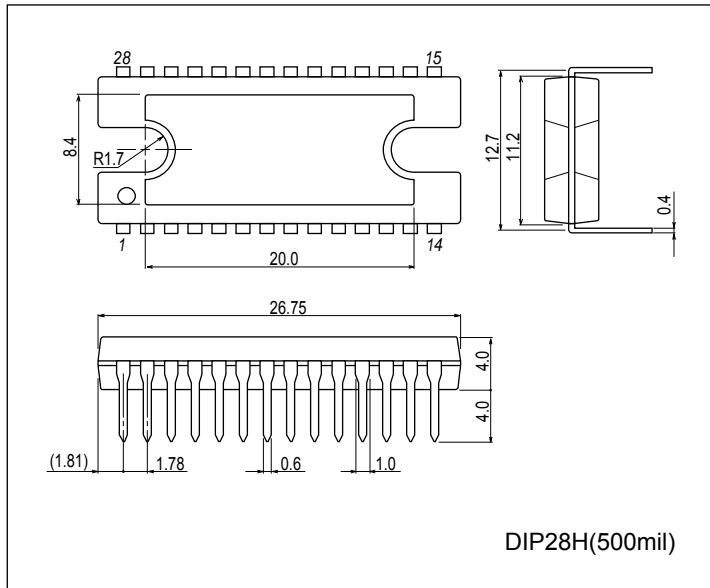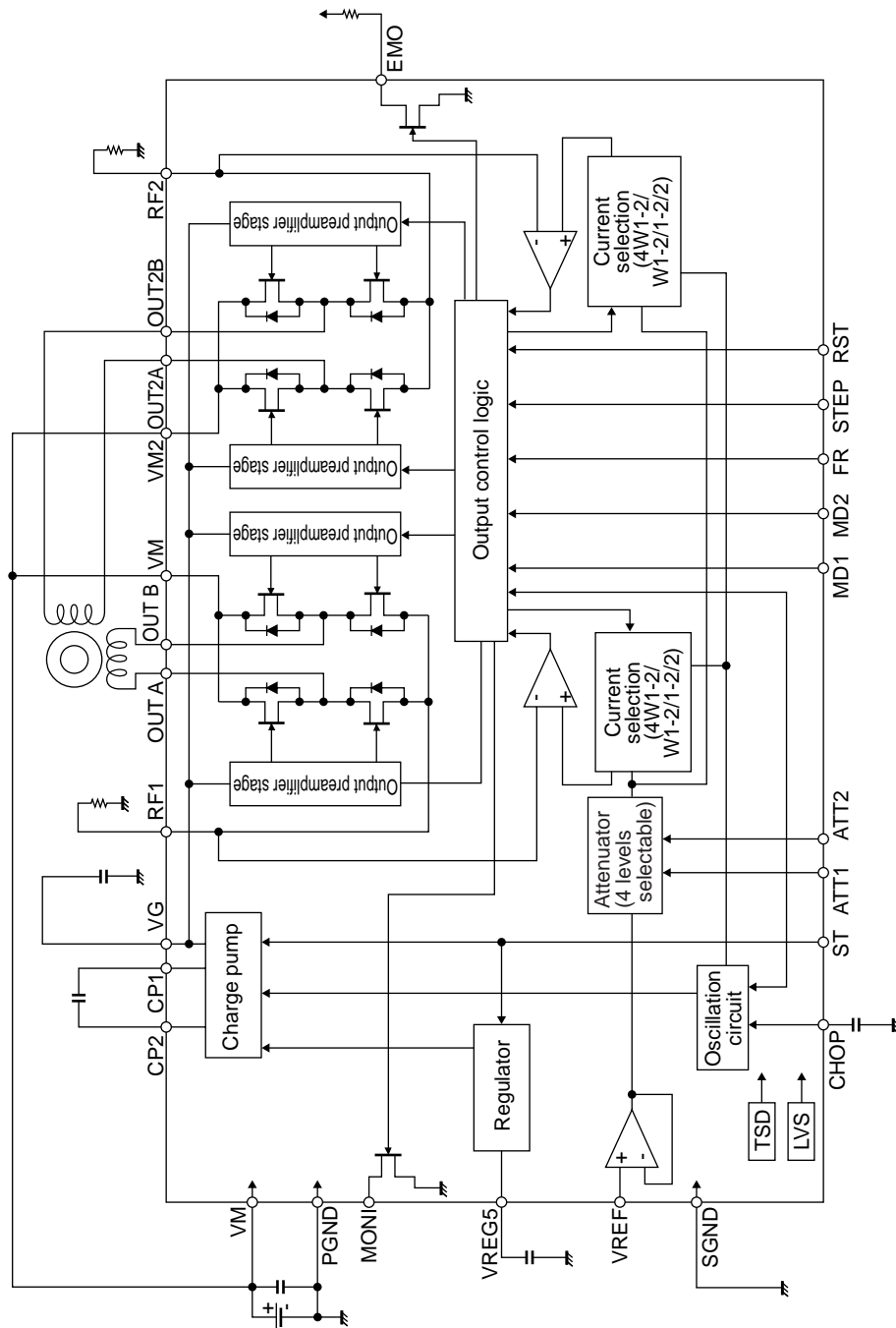
VM	1			28	CP2
VG	2			27	CP1
OUT1A	3			26	VREG5
PGND	4			25	ATT2
VM1	5			24	ATT1
RF1	6			23	EMO
OUT1B	7	**LV8772**		22	CHOP
OUT2A	8			21	MONI
RF2	9			20	RST
VM2	10			19	STEP
PGND	11			18	FR
OUT2B	12			17	MD2
GND	13			16	MD1
VREF	14			15	ST
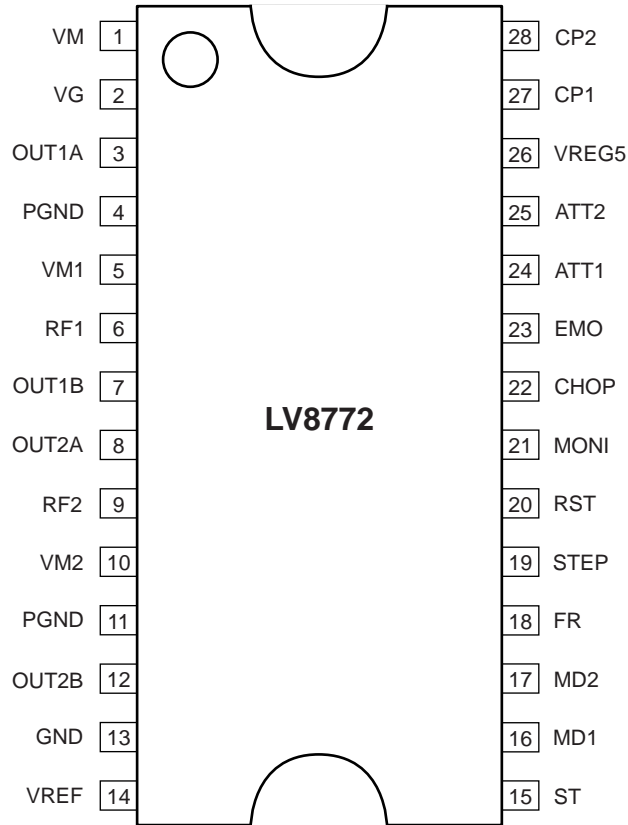
Top view

Pin Functions

Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
25	ATT2	Motor holding current switching pin.	
24	ATT1	Motor holding current switching pin.	
20	RST	RESET input pin	
19	STEP	STEP signal input pin	
18	FR	CW / CCW signal input pin	
17	MD2	Excitation mode switching pin 2	
16	MD1	Excitation mode switching pin 1	
15	ST	Chip enable pin.	

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Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
12	OUT2B	Channel 2 OUTB output pin.	
4/11	PGND	Power system ground.	
10	VM2	Channel 2 motor power supply connection pin.	
9	RF2	Channel 2 current-sense resistor connection pin.	
8	OUT2A	Channel 2 OUTA output pin.	
7	OUT1B	Channel 1 OUTB output pin.	
6	RF1	Channel 1 current-sense resistor connection pin.	
5	VM1	Channel 1 motor power supply pin.	
3	OUT1A	Channel 1 OUTA output pin	
2	VG	Charge pump capacitor connection pin.	
1	VM	Motor power supply connection pin.	
28	CP2	Charge pump capacitor connection pin.	
27	CP1	Charge pump capacitor connection pin	
14	VREF	Constant current control reference voltage input pin	
26	VREG5	Internal power supply capacitor connection pin	

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Pin No.	Pin Name	Pin Functtion	Equivalent Circuit
23	EMO	Output short-circuit state warning output pin.	
21	MONI	Position detection monitor pin.	
22	CHOP	Chopping frequency setting capacitor connection pin	

Description of operation

Input Pin Function

(1) Chip enable function

This IC is switched between standby and operating mode by setting the ST pin. In standby mode, the IC is set to power-save mode and all logic is reset. In addition, the internal regulator circuit and charge pump circuit do not operate in standby mode.

ST	Mode	Internal regulator	Charge pump
Low or Open	Standby mode	Standby	Standby
High	Operating mode	Operating	Operating

Stepping mode drive method

(1) STEP pin function

Input		Operating mode
ST	STP	
Low	*	Standby mode
High	⤴	Excitation step proceeds
High	⤵	Excitation step is kept

(2) Excitation mode setting function

MD1	MD2	Excitation mode	Initial position	
			Channel 1	Channel 2
Low	Low	2 phase excitation	100%	-100%
High	Low	1-2 phase excitation	100%	0%
Low	High	W1-2 phase excitation	100%	0%
High	High	4W1-2 phase excitation	100%	0%

This is the initial position of each excitation mode in the initial state after power-on and when the counter is reset.

(3) Position detection monitoring function

The MONI position detection monitoring pin is of an open drian type.

When the excitation position is in the initial position, the MONI output is placed in the ON state.

(Refer to "Examples of current waveforms in each of the excitation modes.")

(4) Setting constant-current control reference current

This IC is designed to automatically exercise PWM constant-current chopping control for the motor current by setting the output current. Based on the voltage input to the VREF pin and the resistance connected between RF and GND, the output current that is subject to the constant-current control is set using the calculation formula below :

$$I_{OUT} = (VREF/5)/RF \text{ resistance}$$

* The above setting is the output current at 100% of each excitation mode.

The voltage input to the VREF pin can be switched to four-step settings depending on the statuses of the two inputs, ATT1 and ATT2. This is effective for reducing power consumption when motor holding current is supplied.

Attenuation function for VREF input voltage

ATT1	ATT2	Current setting reference voltage attenuation ratio
Low	Low	100%
High	Low	80%
Low	High	50%
High	High	20%

The formula used to calculate the output current when using the function for attenuating the VREF input voltage is given below.

$$I_{OUT} = (VREF/5) \times (\text{attenuation ratio})/RF \text{ resistance}$$

Example : At VREF of 1.5V, a reference voltage setting of 100% [(ATT1, ATT2) = (L, L)] and an RF resistance of
0.22Ω, the output current is set as shown below.

$$I_{OUT} = 1.5V/5 \times 100\%/0.22\Omega = 1.36A$$

If, in this state, (ATT1, ATT2) is set to (H, H), IOUT will be as follows :

$$I_{OUT} = 1.36A \times 20\% = 272mA$$

In this way, the output current is attenuated when the motor holding current is supplied so that power can
be conserved.

(5) Blanking period

If, when exercising PWM constant-current chopping control over the motor current, the mode is switched from decay
to charge, the recovery current of the parasitic diode may flow to the current sensing resistance, causing noise to be
carried on the current sensing resistance pin, and this may result in erroneous detection. To prevent this erroneous
detection, a blanking period is provided to prevent the noise occurring during mode switching from being received.
During this period, the mode is not switched from charge to decay even if noise is carried on the current sensing
resistance pin.
This IC is the blanking time is fixed at approximately 1μs.

(6) Reset function

RST	Operating mode
Low	Normal operation
High	Reset state
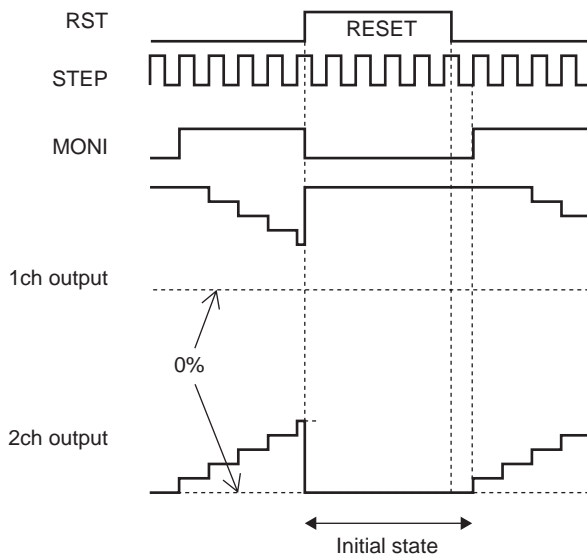
When the RST pin is set to High, the excitation position of the output is forcibly set to the initial state, and the MONI
output is placed in the ON state. When RST is then set to Low, the excitation position is advanced by the next STEP
input.

(7) Forward/reverse switching function

FR	Operating mode
Low	Clockwise (CW)
High	Counter-clockwise (CCW)
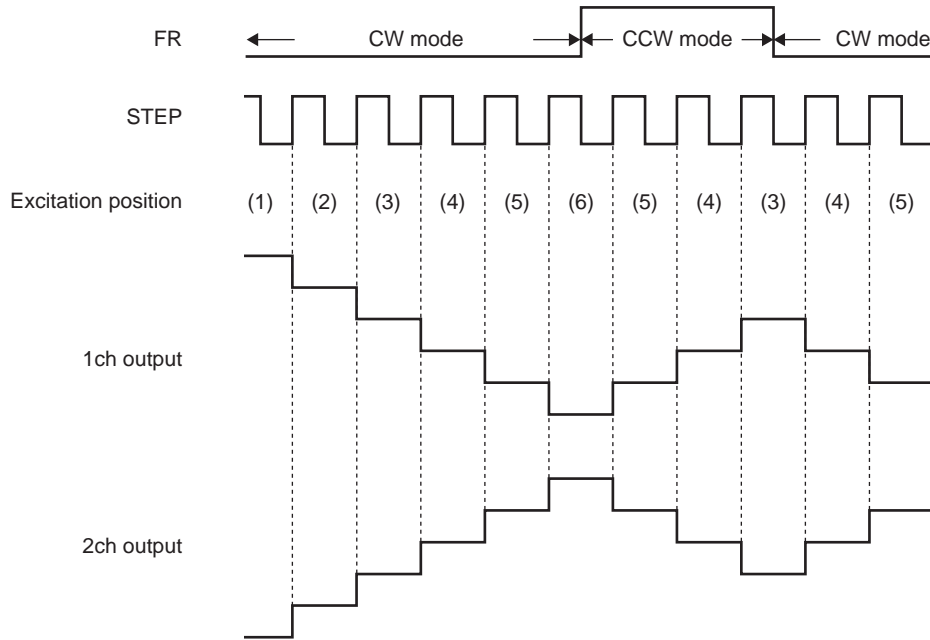
The internal D/A converter proceeds by one bit at the rising edge of the input STEP pulse.

In addition, CW and CCW mode are switched by setting the FR pin.

In CW mode, the channel 2 current phase is delayed by 90° relative to the channel 1 current.

In CCW mode, the channel 2 current phase is advanced by 90° relative to the channel 1 current.

(8) Chopping frequency setting

For constant-current control, this IC performs chopping operations at the frequency determined by the capacitor (Cchop) connected between the CHOP pin and GND.

The chopping frequency is set as shown below by the capacitor (Cchop) connected between the CHOP pin and GND.

$$Fchop = Ichop/ (Cchop \times Vtchop \times 2) \ (Hz)$$

Ichop : Capacitor charge/discharge current, typ 10μA

Vtchop : Charge/discharge hysteresis voltage (Vtup-Vtdown), typ 0.5V

For instance, when Cchop is 180pF, the chopping frequency will be as follows :

$$Fchop = 10μA/ (180pF \times 0.5V \times 2) = 55kHz$$

(9) Output current vector locus (one step is normalized to 90 degrees)
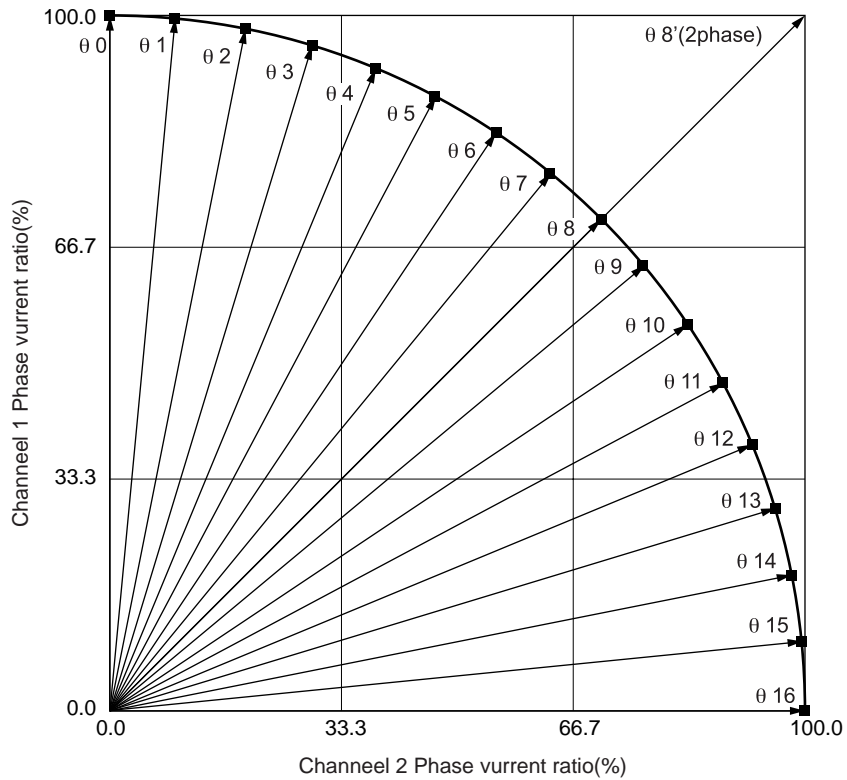
Setting current ration in each excitation mode

STEP	4W1-2 phase (%)		W1-2 phase (%)		1-2 phase (%)		2-phase (%)	
	Channel 1	Channel 2	Channel 1	Channel 2	Channel 1	Channel 2	Channel 1	Channel 2
θ0	100	0	100	0	100	0		
θ1	100	10						
θ2	98	20						
θ3	96	29						
θ4	92	38	92	38				
θ5	88	47						
θ6	83	55						
θ7	77	63						
θ8	70	70	70	70	70	70	100	100
θ9	63	77						
θ10	55	83						
θ11	47	88						
θ12	38	92	38	92				
θ13	29	96						
θ14	20	98						
θ15	10	100						
θ16	0	100	0	100	0	100		

(10) Typical current waveform in each excitation mode

2-phase excitation (CW mode)

1-2 phase excitation (CW mode)

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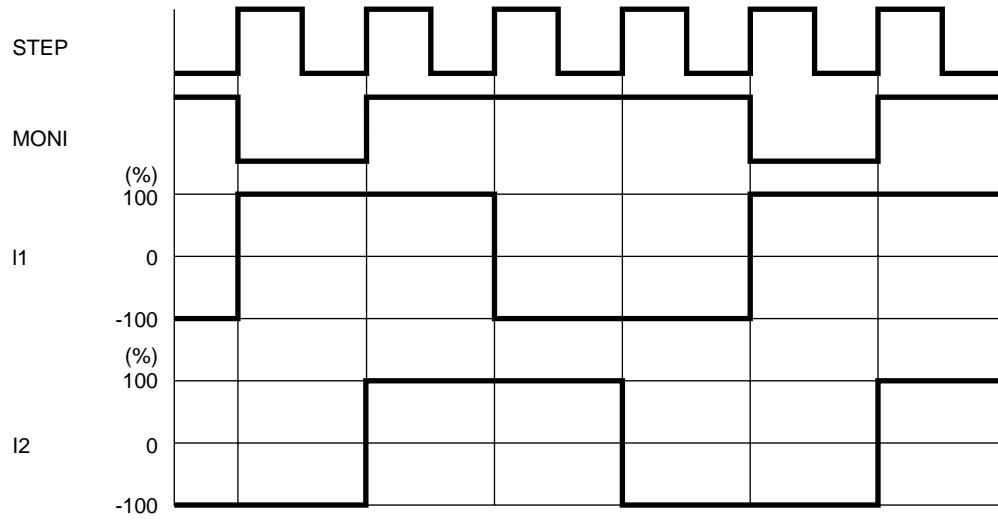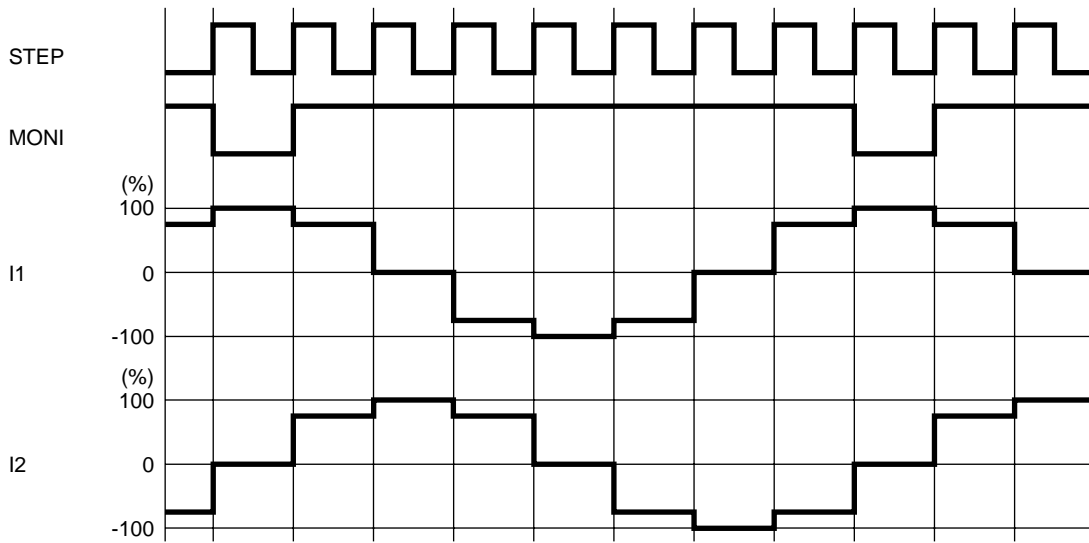
W1-2 phase excitation (CW mode)
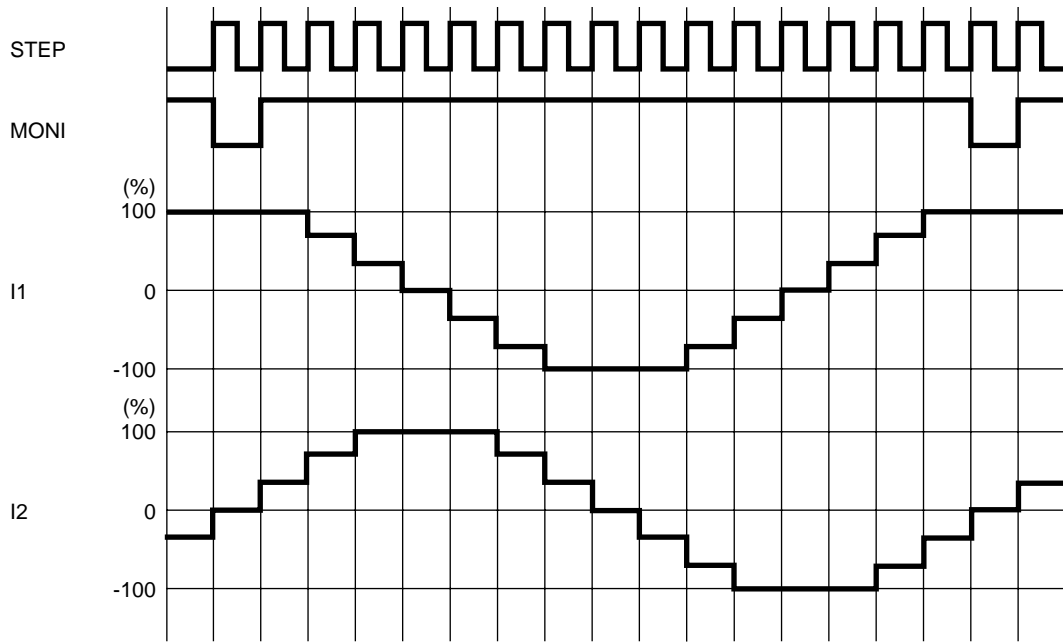
4W1-2 phase excitation (CW mode)
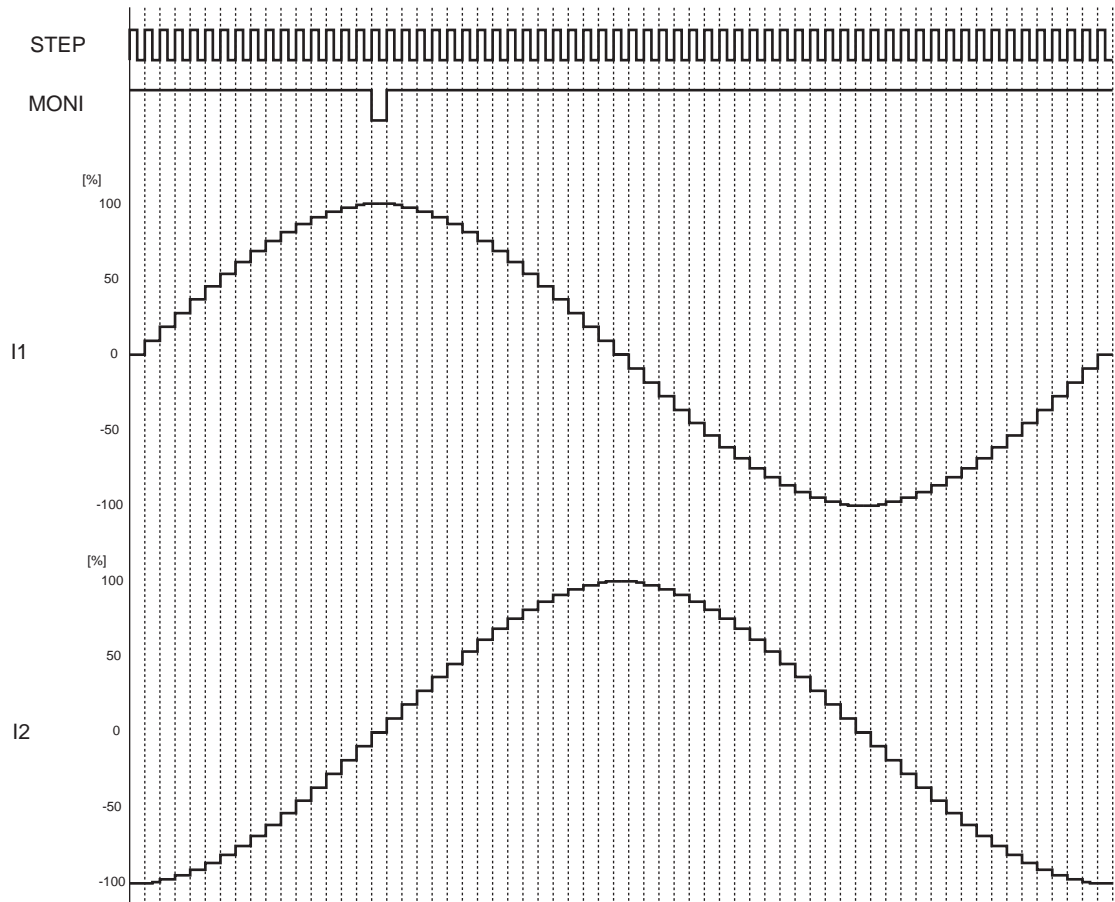
(11) Current control operation specification

(Sine wave increasing direction)

STEP

Coil current

Set current

Forced CHARGE section

Set current

fchop

Current mode　CHARGE　　SLOW　　FAST　　　　　　　CHARGE　　SLOW　FAST
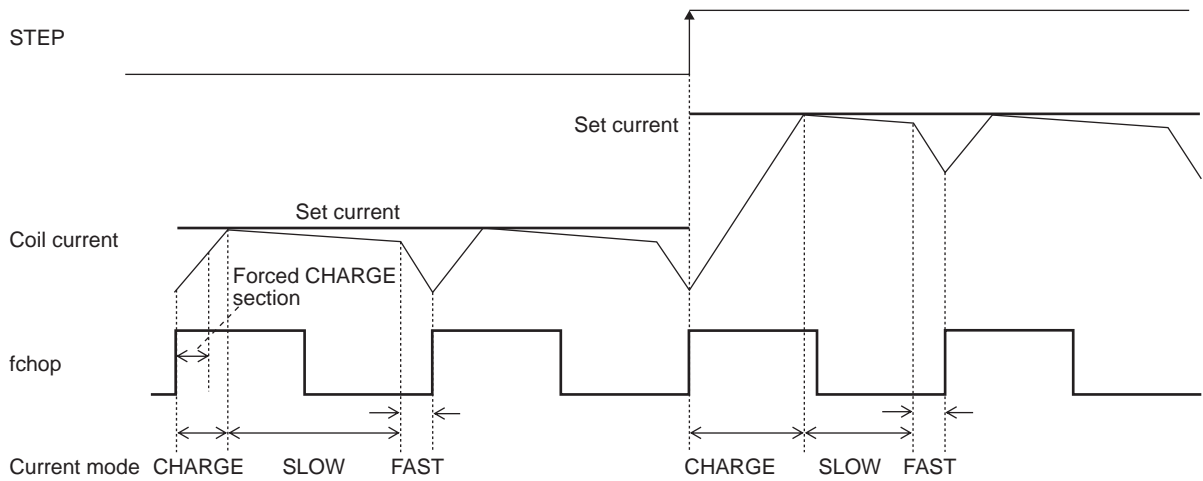
(Sine wave decreasing direction)

STEP

Set current

Coil current

Forced CHARGE section

Set current

fchop

Current mode　CHARGE　　SLOW　　FAST　　　Forced CHARGE　FAST　　CHARGE　　SLOW
　　　　　　　　　　　　　　　　　　　　　　　　section
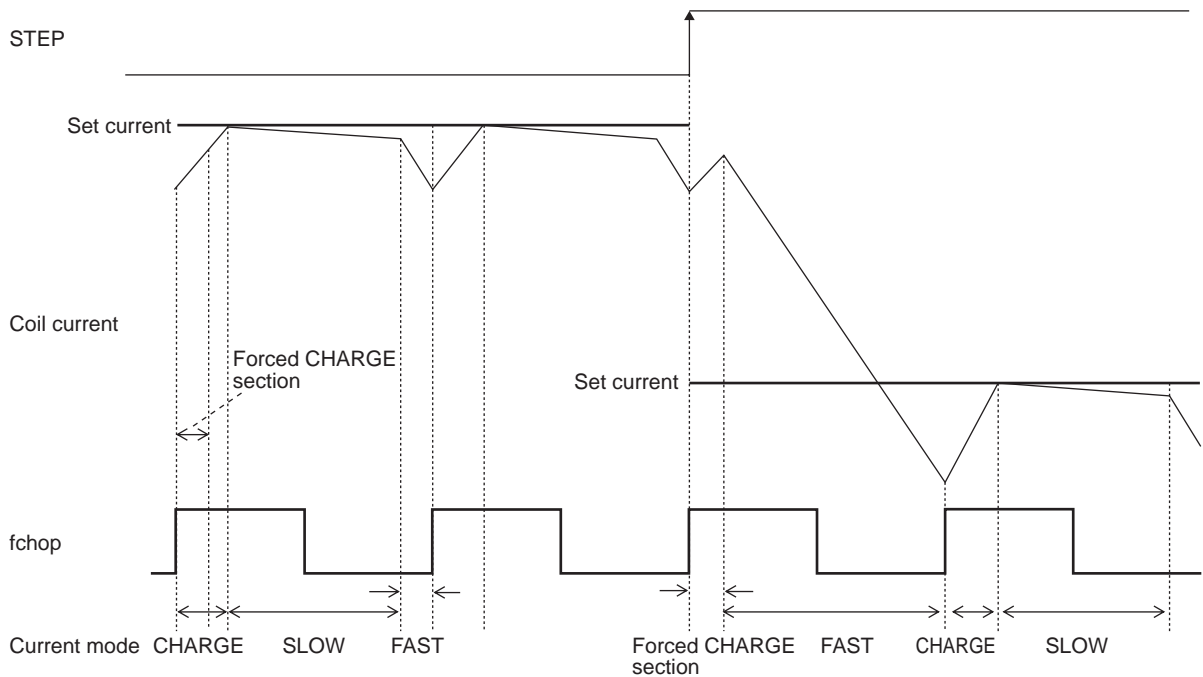
In each current mode, the operation sequence is as described below :
- • At rise of chopping frequency, the CHARGE mode begins. (In the time defined as the "blanking time," the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF).)
- • The coil current (ICOIL) and set current (IREF) are compared in this blanking time.
    When (ICOIL < IREF) state exists ;
        The CHARGE mode up to ICOIL ≥ IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for approximately 1μs.
    When (ICOIL < IREF) state does not exist ;
        The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.

Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

Output short-circuit protection function

This IC incorporates an output short-circuit protection circuit that, when the output has been shorted by an event such as shorting to power or shorting to ground, sets the output to the standby mode and turns on the warning output in order to prevent the IC from being damaged.

This function sets the output to the standby mode for both channels by detecting the short-circuiting in one of the channels.

(1) Output short-circuit protection method

The output short-circuit protection method of LV8772 is a latch method to turn off the output when the output current exceeds the detection current, and to maintain the state.

The detection of the output short-circuited state by the IC causes the output short-circuit protection circuit to be activated.

All the outputs of correspondence ch side where the short-circuit was first detected are switched to the standby mode when the short-circuit is the consecutive between internal timers (approximately 4μs), and the state is maintained. This state is released by setting ST to low.
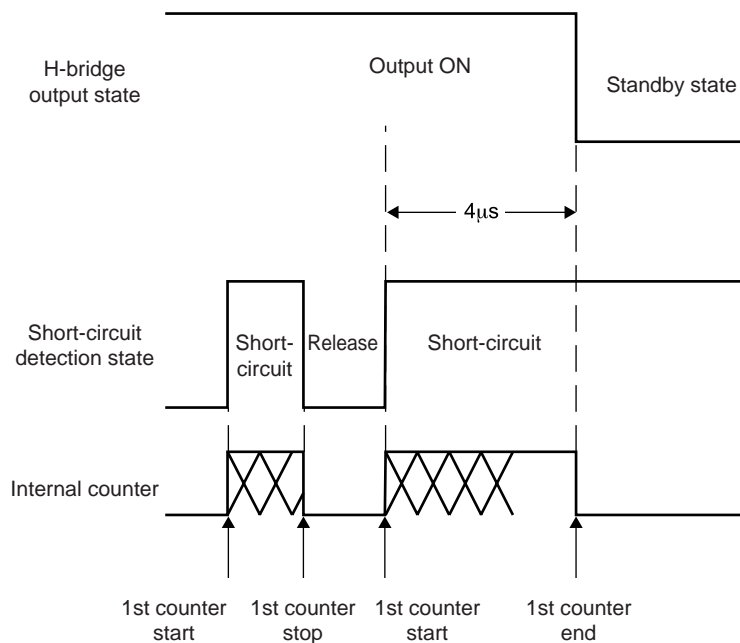
(2) Unusual condition warning output pins (EMO)

The LV8772 is provided with the EMO pin which notifies the CPU of an unusual condition if the protection circuit operates by detecting an unusual condition of the IC. This pin is of the open-drain output type and when an unusual condition is detected, the EMO output is placed in the ON (EMO = Low) state.

Furthermore, the EMO pin is placed in the ON state when one of the following conditions occurs.

1. Shorting-to-power, shorting-to-ground, or shorting-to-load occurs at the output pin and the output short-circuit protection circuit is activated.
2. The IC junction temperature rises and the thermal protection circuit is activated.

Charge Pump Circuit

When the ST pin is set High, the charge pump circuit operates and the VG pin voltage is boosted from the VM voltage to the VM + VREG5 voltage. If the VG pin voltage is not boosted sufficiently, the output cannot be controlled, so be sure to provide a wait time of tONG or more after setting the ST pin High before starting to drive the motor.

ST

VG pin voltage

VM+VREG5

VM+4V

VM

tONG
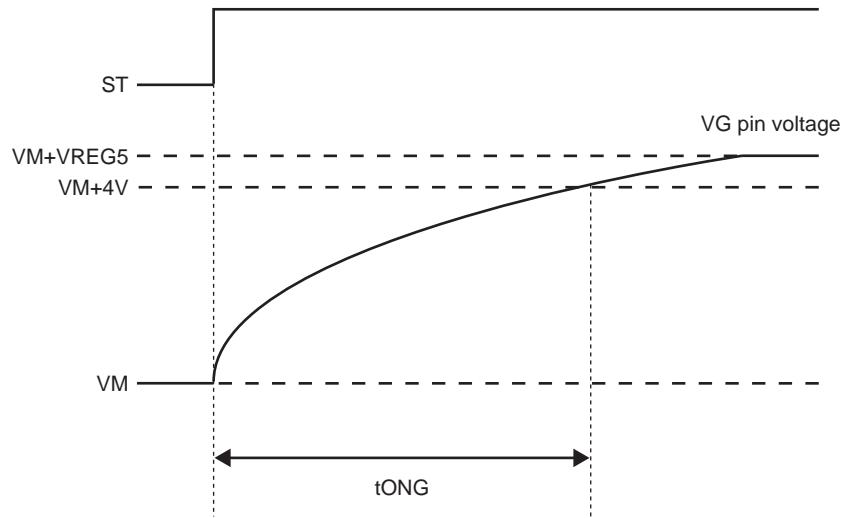
VG Pin Voltage Schematic View

Application Circuit Example

• Stepping motor driver circuit
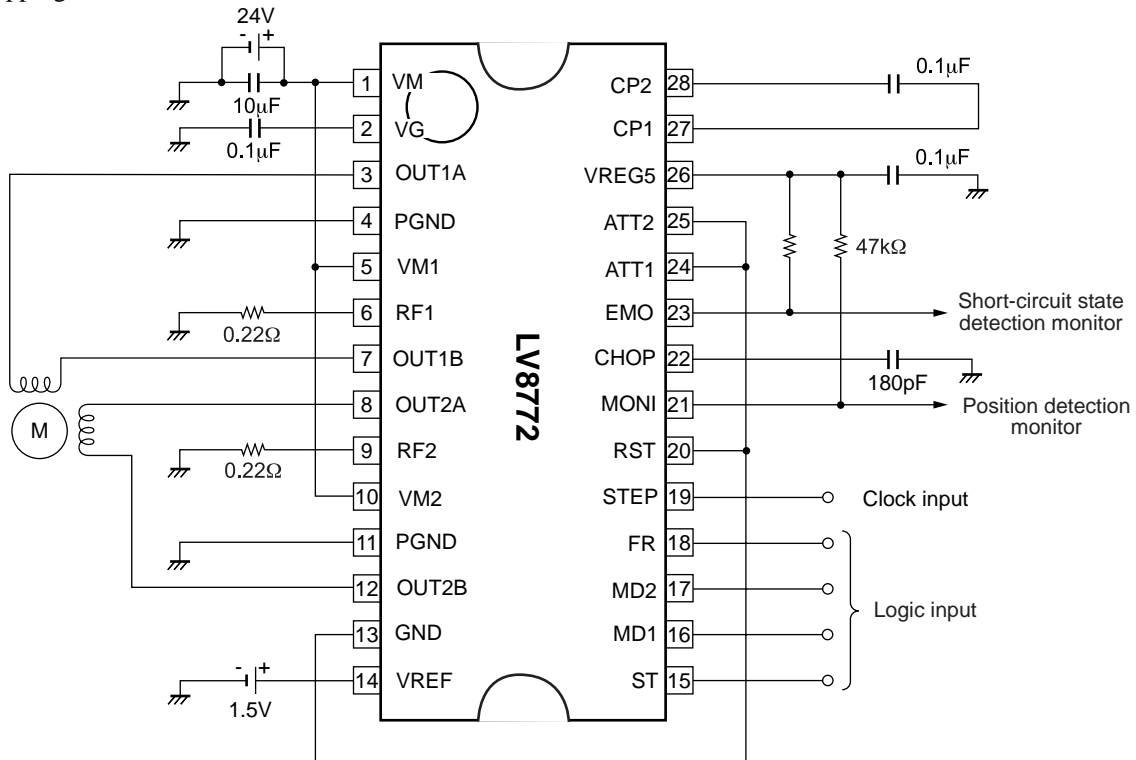
The formulae for setting the constants in the examples of the application circuits above are as follows :

Constant current (100%) setting

When VREF = 1.5V

$$I_{OUT} = VREF/5/RF \text{ resistance}$$
$$= 1.5V/5/0.22\Omega = 1.36A$$

Chopping frequency setting

$$Fchop = Ichop/ (Cchop \times Vtchop \times 2)$$
$$= 10\mu A/ (180pF \times 0.5V \times 2) = 55kHz$$